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# Blind carrier frequency offset estimation for interleaved orthogonal frequency division multiple access uplink with multi-antenna receiver: algorithms and performance analysis

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Abstract: In this study, for interleaved orthogonal frequency division multiple access uplink with multi-antenna receiver, we propose two generalised carrier frequency offset estimators which, respectively, exploit the subspace theory and maximum-likelihood (ML) criterion. We find that, as long as the numbers of multipaths from two of the users are smaller than the number of antennas at the receiver, both the proposed estimators support fully loaded transmissions. We also derive the theoretical performance lower bound for the proposed ML estimator. The numerical results are then provided, which corroborate the proposed studies.

#### 1 Introduction

Orthogonal frequency division multiple access (OFDMA) has recently received a considerable amount of interest [1, 2]. It is widely recognised that OFDMA inherits from orthogonal frequency division multiplexing (OFDM) the weakness of sensitivity to the effect of carrier frequency offset (CFO). The system performance of OFDMA relies heavily on the proper CFO estimation and compensation.

In OFDMA downlink transmissions, a single CFO exists between each transceiver pair, which makes the existing CFO estimators for OFDM directly applicable [3, 4]. However, for OFDMA uplink, multiple CFOs co-exist at the receiver, making the CFO estimation much more challenging. To perform the CFO estimation, various kinds of data-aided estimators [5, 6] as well as the blind estimators [7–15] have been developed. For example, by using the multiple signal classification technique [16], a frequency estimation scheme for interleaved OFDMA uplink that exploits the periodic structure of the signals from each user has been presented in [9]. Based on the observation of [9], several advancements have been proposed later [10–14]. However, [9–14] need null subcarriers or a longer cyclic prefix (CP) to build the noise space which decreases the bandwidth efficiency.

Recently, by exploiting the multi-antenna redundancy at the receiver, Zhang *et al.* [15] proposed two estimators for the interleaved OFDMA uplink based on the subspace theory and maximum-likelihood (ML) criterion, respectively. It was shown in [15] that the two estimators [15] can support fully loaded transmission, which provides higher bandwidth efficiency as compared with the estimators [9-14]. However, both the two estimators in [15] require that the number of antennas at the receiver should be larger than the maximum number of multipaths from all users, which severely limits their applicable scenarios.

In this paper, inspired by the work of [15], we further consider the more general scenarios that the number of multipaths from some users may be larger than the number of antennas at the receiver, and propose two generalised blind CFO estimators for the interleaved OFDMA uplink with multi-antenna receiver. The main contributions of this paper can be summarised as follows:

Two generalised CFO estimators which, respectively, exploit the subspace theory and ML criterion are developed, referred to as 'GSSE' and 'GMLE', respectively. We find that, as long as the numbers of multipaths from two of the users are smaller than the number of antennas at the receiver, both the proposed estimators support fully loaded transmissions. This greatly relaxes the antenna number requirement at the receiver as compared with the work in [15].
We derive the theoretical performance lower bound for the proposed GMLE. The numerical results are also provided, which demonstrate that GMLE behaves better than GSSE and almost achieves the analytical lower bound.

The rest of this paper is organised as follows. Section 2 formulates the problem. The proposed GSSE and GMLE are developed in Sections 3 and 4, respectively. The theoretical performance analysis is presented in Section 5.

Simulation results are given in Section 6 and conclusions are drawn in Section 7.

*Notations:* superscripts  $(\cdot)^*$ ,  $(\cdot)^T$ ,  $(\cdot)^H$ ,  $[\cdot]^\dagger$  and  $E[\cdot]$  represent conjugate, transpose, Hermitian, pseudo inverse and expectation, respectively;  $\mathbf{j} = \sqrt{-1}$  is the imaginary unit;  $||\mathbf{X}||$  denotes the Frobenius norm of  $\mathbf{X}$ , and diag $(\cdot)$  is a diagonal matrix with main diagonal  $(\cdot)$ ; blkdiag $(\cdot)$  represents the block-diagonal matrix operator;  $\mathbb{C}^{m \times n}$  defines the vector space of all  $m \times n$  complex matrices;  $\mathbf{I}_N$  is the  $N \times N$  identity matrix; 0 represents an all-zero matrix with appropriate dimension;  $\otimes$  stands for the Kronecker product;  $\circ$  denotes the element-wise product;  $\text{Tr}\{\cdot\}$  denotes the trace operation; MATLAB matrix representations are adopted, for example,  $X(r_1:r_2, c_1:c_2)$  denotes the submatrix of X with the rows from  $r_1$  to  $r_2$  and the columns from  $c_1$  to  $c_2$ .

#### 2 Problem formulation

Consider a multiuser OFDMA system with K users, N subcarriers and M antennas at the receiver. All subcarriers are sequentially indexed with  $\{i\}$ , i=0, 1, ..., N-1, and are equally divided into Q subchannels, each having P=N/Q subcarriers. The qth subchannel consists of subcarriers with index set of  $\{q, Q+q, ..., (P-1)Q+q\}, q=0, 1, ..., Q-1$ . Each subchannel will be exclusively assigned to one user, and thus, no subchannel can be shared by more than one user. To ease the presentation, we assume the system is fully loaded in this paper, that is, K=Q.

Denote the normalised CFO of the *k*th user as  $\xi^{(k)} = \Delta f^{(k)} / \Delta f$ , where  $\Delta f$  is the subcarrier spacing and  $\Delta f^{(k)}$  is the real CFO of the *k*th user. We assume  $\xi^{(k)} \in (-0.5, 0.5)$ . The channel impulse response from the *k*th user to the *m*th antenna can be modelled as  $h_m^{(k)}(\tau) = \sum_{l=1}^{L^{(k)}} h_{l,m}^{(k)} \delta\left(\tau - \tau_l^{(k)}\right)$  where  $L^{(k)}$  is the number of multipaths, whereas  $h_{l,m}^{(k)}$  and  $\tau_l^{(k)}$  are the complex amplitude and delay for the *l*th multipath, respectively. We assume  $h_{l,m}^{(k)}$ 's are independent and identically distributed (i.i.d.) complex Gaussian variables with zero mean and power  $E\left\{|h_{l,m}^{(k)}|^2\right\} = 1/L^{(k)}$  such that the total power is normalised, that is,  $E\left\{\sum_{l=1}^{L^{(k)}} \left|h_{l,m}^{(k)}\right|^2\right\} = 1$ . Assume that the delays are rounded to the nearest sampling position and  $\tau_l^{(k)}$ 's are of integer values. We also consider the fading channels are constant over successive  $L_s$  block durations in this paper.

We define the user index set  $C = \{c_1, c_2, \ldots, c_{N_C}\}$ , where  $L^{(c_i)} < M$ ,  $i = 1, 2, \ldots, N_C$ , and  $N_C$  denotes the number of users whose multipath numbers are less than M. In this paper, we assume  $N_C \ge 2$ , that is, there are at least two users whose multipath numbers are less than M. Here, we should note that the work in [15] requires that the multipath numbers of all users should be less than M. Thus, the antenna number requirement at the receiver in this paper is much more relaxed as compared with [15].

Assume the *k*th user occupies the  $q^{(k)}$ th subchannel and let  $s_g^{(k)} = \begin{bmatrix} s_{0,g}^{(k)}, s_{1,g}^{(k)}, \dots, s_{P-1,g}^{(k)} \end{bmatrix}^T$  be the *P* information symbols of the *k*th user in the *g*th OFDMA block. Denote  $\chi_g^{(k)} = e^{j(2\pi(g-1)(N+G)\xi^{(k)}/N)}$  as the phase shift of the *k*th user accumulated from the previous g-1 blocks with *G* being the length of CP. Then, in noise-free environment, the received signal at the *m*th receiving antenna element from the *K* users after CP removal can be expressed as [15]

where the expression in the bracket stands for the frequency-domain channel response at the  $(pK+q^{(k)})$ th subcarrier from the *k*th user to the *m*th antenna, and (see (1))

We further define (see equation at the bottom of the page)

where  $L_{\text{sum}} = \sum_{k=1}^{K} L^{(k)}$ . Stacking the received signals from all M antennas, we obtain the following space-domain snapshot with the presence of noise

$$\gamma_{n,g} = [\gamma_{1,g}(n), \ \gamma_{2,g}(n), \ \dots, \ \gamma_{M,g}(n)]^{1}$$
  
=  $H z_{n,g} + n_{n,g}$  (2)

where  $\mathbf{n}_{n,g}$  is a length-*M* additive white Gaussian noise (AWGN) vector with covariance matrix  $\sigma_n^2 \mathbf{I}_M$  at the *n*th sample in the *g*th OFDMA block, and

$$\mathbf{z}_{n,g} = \left[ \left( \mathbf{z}_{n,g}^{(1)} \right)^{\mathrm{T}}, \left( \mathbf{z}_{n,g}^{(2)} \right)^{\mathrm{T}}, \dots, \left( \mathbf{z}_{n,g}^{(K)} \right)^{\mathrm{T}} \right]^{\mathrm{T}} \in \mathbb{C}^{L_{\mathrm{sum}} \times 1}$$
$$\mathbf{z}_{n,g}^{(k)} = \left[ z_{1,n,g}^{(k)}, z_{2,n,g}^{(k)}, \dots, z_{L^{(k)},n,g}^{(k)} \right]^{\mathrm{T}} \in \mathbb{C}^{L^{(k)} \times 1}$$

#### 3 Proposed GSSE

By defining

$$\bar{z}_{p,g}^{(k)} = \begin{cases} z_{p,g}^{(k)}, & L^{(k)} < M \\ H^{(k)} z_{p,g}^{(k)}, & L^{(k)} \ge M \end{cases}$$
(3)

$$\gamma_{m,g}^{(k)}(n) = \sum_{k=1}^{K} \frac{\chi_g^{(k)}}{\sqrt{N}} \sum_{p=0}^{P-1} \left( \sum_{l=1}^{L^{(k)}} h_{l,m}^{(k)} \mathrm{e}^{-\mathbf{j} \left( 2\pi (pK + q^{(k)}) \tau_l^{(k)} / N \right)} \right) s_{p,g}^{(k)} \mathrm{e}^{\mathbf{j} \left( \left( 2\pi (pK + q^{(k)} + \xi^{(k)}) / N \right) n \right)}$$

$$= \sum_{k=1}^{K} \sum_{l=1}^{L^{(k)}} h_{l,m}^{(k)} z_{l,n,g}^{(k)}$$

$$(1)$$

$$\boldsymbol{h}_{l}^{(k)} = \begin{bmatrix} h_{l,1}^{(k)}, h_{l,2}^{(k)}, \dots, h_{l,M}^{(k)} \end{bmatrix}^{\mathrm{T}} \in \mathbb{C}^{M \times 1}, \quad \boldsymbol{H}^{(k)} = \begin{bmatrix} \boldsymbol{h}_{1}^{(k)}, \boldsymbol{h}_{2}^{(k)}, \dots, \boldsymbol{h}_{L^{(k)}}^{(k)} \end{bmatrix} \in \mathbb{C}^{M \times L^{(k)}}$$
$$\boldsymbol{H} = \begin{bmatrix} \boldsymbol{H}^{(1)}, \boldsymbol{H}^{(2)}, \dots, \boldsymbol{H}^{(K)} \end{bmatrix} \in \mathbb{C}^{M \times L_{\mathrm{sum}}}$$

we can rewrite (2) into

$$\gamma_{p,g} = \sum_{k=1}^{K} H^{(k)} z_{p,g}^{(k)} + n_{p,g}$$
  
=  $\sum_{k' \in \mathcal{C}} H^{(k')} \bar{z}_{p,g}^{(k')} + \sum_{k \notin \mathcal{C}} I_M \bar{z}_{p,g}^{(k)} + n_{p,g}$  (4)

Treat  $\theta^{(k)} = (q^{(k)} + \xi^{(k)})/K$  as the effective CFO of the *k*th user. For t = 0, 1, ..., K-1, there holds

$$\bar{\boldsymbol{z}}_{p+t^{p},g}^{(k)} = \mathrm{e}^{\mathbf{j}2\pi t \theta^{(k)}} \bar{\boldsymbol{z}}_{p,g}^{(k)}$$
(5)

and

$$\gamma_{p+tP,g} = \sum_{k' \in \mathcal{C}} e^{\mathbf{j}2\pi t \theta^{(k')}} \boldsymbol{H}^{(k')} \bar{\boldsymbol{z}}_{p,g}^{(k')} + \sum_{k \notin \mathcal{C}} e^{\mathbf{j}2\pi t \theta^{(k)}} \boldsymbol{I}_M \bar{\boldsymbol{z}}_{p,g}^{(k)} + \boldsymbol{n}_{p+tP,g}$$
(6)

Stacking space-domain snapshot vectors from K equally spaced time samples (P samples apart) give the following length-MK vector

$$\boldsymbol{a}_{p,g} = \left[\boldsymbol{\gamma}_{p,g}^{\mathrm{T}}, \; \boldsymbol{\gamma}_{p+P,g}^{\mathrm{T}}, \; \dots, \; \boldsymbol{\gamma}_{p+N-P,g}^{\mathrm{T}}\right]^{\mathrm{T}}$$
(7)

Define

$$\boldsymbol{v}^{(k)} = \begin{bmatrix} 1, \ \mathbf{e}^{\mathbf{j}2\pi\theta^{(k)}}, \ \dots, \ \mathbf{e}^{\mathbf{j}2\pi(K-1)\theta^{(k)}} \end{bmatrix}^{\mathrm{T}} \in \mathbb{C}^{K\times 1}$$
(8)

$$\boldsymbol{V}^{(k)} = \begin{cases} \boldsymbol{v}^{(k)} \otimes \boldsymbol{H}^{(k)}, & L^{(k)} < M\\ \boldsymbol{v}^{(k)} \otimes \boldsymbol{I}_M, & L^{(k)} \ge M \end{cases}$$
(9)

Let  $\bar{L}^{(k)}$  equal  $L^{(k)}$  if  $L^{(k)} < M$  and M elsewhere, that is,  $\bar{L}^{(k)} = \min\{L^{(k)}, M\}$ . Then, we have  $V^{(k)} \in \mathbb{C}^{KM \times \bar{L}^{(k)}}$  and  $\bar{z}_{p,g}^{(k)} \in \mathbb{C}^{\bar{L}^{(k)} \times 1}$ . Denote  $\bar{L}_{sum} = \sum_{k=1}^{K} \bar{L}^{(k)}$ . From (6), we can further rewrite (7) into the following matrix-form

$$\boldsymbol{a}_{p,g} = \bar{\boldsymbol{V}}\boldsymbol{z}_{p,g} + \bar{\boldsymbol{n}}_{p,g} \tag{10}$$

where

$$\boldsymbol{V} = [\boldsymbol{V}^{(1)}, \ \boldsymbol{V}^{(2)}, \ \dots, \ \boldsymbol{V}^{(K)}] \in \mathbb{C}^{MK \times \bar{L}_{\text{sum}}}$$
$$\bar{\boldsymbol{z}}_{p,g} = \left[ \left( \bar{\boldsymbol{z}}_{p,g}^{(1)} \right)^{\text{T}}, \left( \bar{\boldsymbol{z}}_{p,g}^{(2)} \right)^{\text{T}}, \ \dots, \ \left( \bar{\boldsymbol{z}}_{p,g}^{(K)} \right)^{\text{T}} \right]^{\text{T}} \in \bar{L}_{\text{sum}} \times 1$$
$$\bar{\boldsymbol{n}}_{p,g} = \left[ \boldsymbol{n}_{p,g}^{\text{T}}, \ \boldsymbol{n}_{p+P,g}^{\text{T}}, \ \dots, \ \boldsymbol{n}_{p+N-P,g}^{\text{T}} \right]^{\text{T}} \in \mathbb{C}^{MK \times 1}$$

The correlation matrix of  $a_{p,g}$  can be computed as

$$\boldsymbol{R}_{i} = E\left\{\boldsymbol{a}_{p,g}\boldsymbol{a}_{p,g}^{\mathrm{H}}\right\} = \boldsymbol{V}\boldsymbol{R}_{\overline{z}\overline{z}}\boldsymbol{V}^{\mathrm{H}} + \sigma_{n}^{2}\boldsymbol{I}_{MK}$$
(11)

where  $\boldsymbol{R}_{\bar{z}\bar{z}} = E\left\{\bar{\boldsymbol{z}}_{p,g}\bar{\boldsymbol{z}}_{p,g}^{\mathrm{H}}\right\}$ . Assume  $L^{(k)} \leq P$  and  $\tau_{L^{(k)}}^{(k)} < P$ . Following the similar steps of Appendix B in [15], we obtain

$$\boldsymbol{R}_{\overline{z}\overline{z}} = \text{blkdiag}\left\{\boldsymbol{R}_{\overline{z}\overline{z}}^{(1)}, \ \boldsymbol{R}_{\overline{z}\overline{z}}^{(2)}, \ \dots, \ \boldsymbol{R}_{\overline{z}\overline{z}}^{(K)}\right\}$$
(12)

where

$$\boldsymbol{R}_{\overline{z}\overline{z}}^{(k)} = \begin{cases} \frac{1}{K} \boldsymbol{I}_{L^{(k)}}, & L^{(k)} < M \\ \frac{1}{K} \boldsymbol{H}^{(k)} (\boldsymbol{H}^{(k)})^{\mathrm{H}}, & L^{(k)} \ge M \end{cases}$$

Owing to the random property of the wireless channels, the non-singularity of  $R_{\overline{zz}}$  can be readily guaranteed.

Perform singular value decomposition (SVD) of  $R_i$ , which gives

$$\boldsymbol{R}_{i} = \begin{bmatrix} \boldsymbol{U}_{s}, \ \boldsymbol{U}_{n} \end{bmatrix} \boldsymbol{\Sigma}_{a} \begin{bmatrix} \boldsymbol{U}_{s}, \ \boldsymbol{U}_{n} \end{bmatrix}^{\mathrm{H}}$$
(13)

where  $U_s \in \mathbb{C}^{MK \times \bar{L}_{sum}}$  and  $U_n \in \mathbb{C}^{MK \times (MK - \bar{L}_{sum})}$  represent the signal and the noise space matrices, respectively.

We define the following parameterised Vandermonde vector  $\mathbb{B}^{(k)}(\xi)$  with respect to  $\xi$  as

$$\mathbb{B}^{(k)}(\xi) = \left[1, \ e^{\mathbf{j}\left(2\pi(q^{(k)}+\xi)/K\right)}, \ \dots, \ e^{\mathbf{j}\left(2\pi(K-1)(q^{(k)}+\xi)/K\right)}\right]^{\mathrm{T}}$$
(14)

where  $\xi \in (-0.5, 0.5)$ . The following 'Lemmas' are the key properties to design the generalised subspace-based CFO estimator.

Lemma 1: For the kth user with  $L^{(k)} < M$ , we have the following observations:

• For 'any' non-zero length-*M* vector *x*, there holds

$$\left(\mathbb{B}^{(k)}(\boldsymbol{\xi}^{(k)}) \otimes \boldsymbol{x}\right)^{\mathsf{H}} \boldsymbol{U}_{\boldsymbol{n}} = \begin{cases} = 0, & \boldsymbol{x} \in \operatorname{Span} \{\boldsymbol{H}^{(k)}\} \\ \neq 0, & \boldsymbol{x} \notin \operatorname{Span} \{\boldsymbol{H}^{(k)}\} \end{cases}$$
(15)

• Given 'any'  $M \times L^{(k)}$  matrix X with full column rank and  $\xi \neq \xi^{(k)}$ , there holds

$$\left(\mathbb{B}^{(k)}(\xi)\otimes X\right)^{\mathrm{H}}U_{n}\neq0\tag{16}$$

When  $x \in \text{Span}\{H^{(k)}\},\$ Proof: we know  $\mathbb{B}^{(k)}(\xi^{(k)}) \otimes \mathbf{x} = \mathbf{v}^{(k)} \otimes \mathbf{x}$  belongs to the column space of  $U_s$ . Thus, it is orthogonal to the noise space matrix  $U_n$ . Moreover, when  $x \notin \text{Span}\{H^{(k)}\}$ , according to Appendix C in [15], we know  $\mathbb{B}^{(k)}(\xi^{(k)}) \otimes x$  is linearly independent with the column vectors of  $U_s$ . Thus, it cannot be orthogonal to  $U_n$ . Then, we concentrate on the proof of (16) which can be obtained from the method of contradiction. According to Appendix D in [15], if  $(\mathbb{B}^{(k)}(\xi) \otimes X)^{H} U_{n} = 0$  holds, the column space of  $X \in \mathbb{C}^{M \times L^{(k)}}$  should simultaneously belong to the column space of both  $I_{M}$  and each  $\boldsymbol{H}^{(k')} \in \mathbb{C}^{M \times L^{(k')}}, k' \in \mathcal{C}.$  However, when the number of users whose multipath number is less than M is not less than 2, that is,  $N_{\mathcal{C}} \geq 2$ , this has zero possibility because of the random property of the wireless channels. From this contradiction, we arrive at (16). Here it needs to be mentioned that, when  $N_c = 1$ , that is, there is only one user whose multipath number is less than M, we have the equation  $\left(\mathbb{B}^{(k)}(\xi)\otimes X\right)^{\mathrm{H}}U_{n}=0$  since X can simultaneously

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belong to the column space of  $I_M$  and  $H^{(k)}$ . Hence,  $N_C \ge 2$  is required in this paper to guarantee the validity of Lemma 1.

Lemma 2: For the kth user with  $L^{(k)} \ge M$ , there holds

$$\left(\mathbb{B}^{(k)}(\xi) \otimes \boldsymbol{I}_{M}\right)^{\mathrm{H}} \boldsymbol{U}_{\boldsymbol{n}} \begin{cases} = 0, & \xi = \xi^{(k)} \\ \neq 0, & \xi \neq \xi^{(k)} \end{cases}$$
(17)

Proof: Note that  $\mathbb{B}^{(k)}(\xi^{(k)}) \otimes I_M = v^{(k)} \otimes I_M$  serve as a submatrix of V. Thus, we can readily obtain that it is orthogonal to the noise space matrix  $U_n$ , that is,  $(\mathbb{B}^{(k)}(\xi^{(k)}) \otimes I_M)^H U_n = 0$ . Then, we concentrate on  $(\mathbb{B}^{(k)}(\xi) \otimes I_M)^H U_n \neq 0$  when  $\xi \neq \xi^{(k)}$  which can be proved from the method of contradiction. According to Appendix D in [15], if  $(\mathbb{B}^{(k)}(\xi) \otimes I_M)^H U_n = 0$ ,  $\xi \neq \xi^{(k)}$ , holds, the column space of  $I_M$  should simultaneously belong to the column space of each  $H^{(k')} \in \mathbb{C}^{M \times L^{(k')}}$ ,  $k' \in C$ . However, this is impossible when  $N_C \geq 1$ . From this contradiction, we arrive at  $\mathbb{B}^{(k)}(\xi) \otimes I_M)^H U_n \neq 0$  when  $\xi \neq \xi^{(k)}$ .

From the above two lemmas, we can design the CFO estimator as follows. For any non-zero length-M vector x, we have

$$\left(\mathbb{B}^{(k)}(\xi)\otimes \boldsymbol{x}\right)^{\mathrm{H}}\boldsymbol{U}_{\boldsymbol{n}}\boldsymbol{U}_{\boldsymbol{n}}^{\mathrm{H}}\left(\mathbb{B}^{(k)}(\xi)\otimes \boldsymbol{x}\right)=\boldsymbol{x}^{\mathrm{H}}\Pi^{(k)}(\xi)\boldsymbol{x}\qquad(18)$$

where  $\Pi^{(k)}(\xi) = \left(\mathbb{B}^{(k)}(\xi) \otimes I_M\right)^{\mathrm{H}} U_n U_n^{\mathrm{H}} \left(\mathbb{B}^{(k)}(\xi) \otimes I_M\right).$ 

For the *k*th user with  $L^{(k)} < M$ , Lemma 1 indicates that  $\Pi^{(k)}(\xi)$  drops its rank to  $M-L^{(k)}$  if and only if  $\xi = \xi^{(k)}$ . Thus, the CFO estimate can be obtained similar to [15]. Specifically, for each trial  $\xi$ , we compute the *M* eigenvalues of the matrix  $\Pi^{(k)}(\xi)$  denoted by  $\kappa_1^{(k)}(\xi)$ ,  $\kappa_2^{(k)}(\xi)$ , ...,  $\kappa_M^{(k)}(\xi)$  in ascending order, and then calculate the summation of the smallest  $L^{(k)}$  eigenvalues of  $\Pi^{(k)}(\xi)$  as the cost of current trial value. The CFO estimate for the *k*th user is the trial value that gives the minimum cost, that is

$$\hat{\xi}^{(k)} = \arg\min_{\xi} \sum_{i=1}^{L^{(k)}} \kappa_i^{(k)}(\xi)$$
 (19)

On the other side, for the *k*th user with  $L^{(k)} \ge M$ , according to Lemma 2, we know  $\left(\mathbb{B}^{(k)}(\xi) \otimes I_M\right)^H U_n$  equals zero matrix if and only if  $\xi = \xi^{(k)}$ . Thus, the CFO can be obtained by

$$\hat{\xi}^{(k)} = \arg\min_{\xi} \operatorname{Tr}\left\{\Pi^{(k)}(\xi)\right\}$$
(20)

Note that  $\operatorname{Tr}\{\Pi^{(k)}(\xi)\} = \sum_{i=1}^{M} \kappa_i^{(k)}(\xi)$ . Hence, combining both (19) and (20), the proposed GSSE can be described as the following unified form

$$\hat{\xi}^{(k)} = \arg\min_{\xi} g_k(\xi) \tag{21}$$

where the cost function is expressed as

$$g_k(\xi) = \sum_{i=1}^{\bar{L}^{(k)}} \kappa_i^{(k)}(\xi)$$
(22)

In noise-free environment, for both the situations with  $L^{(k)} <$ 

*M* and  $L^{(k)} \ge M$ , one can imagine that,  $g_k(\xi) = 0$  when  $\xi = \xi^{(k)}$ , whereas  $g_k(\xi) > 0$  when  $\xi \neq \xi^{(k)}$ . Hence, by searching for the minimum point of  $g_k(\xi)$ , we could obtain the estimation for  $\xi^{(k)}$ . Moreover, for the *k*th user with  $L^{(k)} < M$ , it is seen from Lemma 1 that, after deriving CFO,  $(\Pi^{(k)}(\hat{\xi}^{(k)}))^{\perp}$  can be considered as the estimate of  $H^{(k)}$  with certain ambiguity.

#### 4 Proposed GMLE

The parameter set of interest for the *k*th user can be expressed as

$$\omega^{(k)} = \begin{cases} \{\xi^{(k)}, H^{(k)}\} & L^{(k)} < M \\ \xi^{(k)} & L^{(k)} \ge M \end{cases}$$
(23)

The corresponding whole parameter set for the *K* users is given by  $\omega = \{\omega^{(1)}, \omega^{(2)}, ..., \omega^{(K)}\}$ . We rewrite (10) as the following parameterised equation

$$\boldsymbol{a}_{p,g} = \boldsymbol{V}(\omega)\bar{\boldsymbol{z}}_{p,g} + \bar{\boldsymbol{n}}_{p,g}$$
(24)

where

$$V(\omega) = \left[ V^{(1)}(\omega^{(1)}), V^{(2)}(\omega^{(2)}), \dots, V^{(K)}(\omega^{(K)}) \right]$$
$$V^{(k)}(\omega^{(k)}) = \begin{cases} v^{(k)}(\xi^{(k)}) \otimes H^{(k)} & L^{(k)} < M \\ v^{(k)}(\xi^{(k)}) \otimes I_M & L^{(k)} \ge M \end{cases}$$
$$\mathbf{v}^{(k)}(\xi^{(k)}) = \left[ 1, e^{\mathbf{j}2\pi(q^{(k)} + \xi^{(k)}/K)}, \dots, e^{\mathbf{j}2\pi(K-1)(q^{(k)} + \xi^{(k)}/K)} \right]^{\mathrm{T}}$$

Concatenating the vectors  $a_{p,g}$ , p = 0, 1, ..., P-1,  $g = 1, 2, ..., L_s$ , we obtain the following  $MK \times PL_s$  matrix

$$A = \left[A_{g}, A_{g+1}, \dots, A_{g+L_{s}-1}\right] = V(\omega)Z + N$$
 (25)

where  $A_g = [a_{0,g}, a_{1,g}, \dots, a_{P-1,g}]$ , Z and N denote the corresponding signal and AWGN matrices, that is,  $Z = [Z_g, Z_{g+1}, \dots, Z_{g+L_s-1}]$  and  $N = [\bar{N}_g, \bar{N}_{g+1}, \dots, \bar{N}_{g+L_s-1}]$  with  $Z_g = [\bar{z}_0, \bar{z}_{1,g}, \dots, \bar{z}_{P-1,g}]$  and  $\bar{N}_g = [\bar{n}_{0,g}, \bar{n}_{1,g}, \dots, \bar{n}_{P-1,g}]$ .

From (25), the ML parameter estimation of  $\omega$  can be obtained by

$$\hat{\boldsymbol{\omega}} = \arg\max_{\tilde{\boldsymbol{\omega}}} \operatorname{Tr}\left\{\boldsymbol{A}^{\mathrm{H}}\boldsymbol{P}_{\boldsymbol{V}}(\tilde{\boldsymbol{\omega}})\boldsymbol{A}\right\}$$
(26)

where  $\tilde{\mathbf{x}}$  stands for the trial value of  $\mathbf{x}$ , and  $P_V(\tilde{\omega}) = V(\tilde{\omega}) [V(\tilde{\omega})]^{\dagger}$  is the projection operator onto the space spanned by the columns of the matrix  $V(\tilde{\omega})$ . The alternating projection algorithm [17] is then applied to reduce the multi-dimensional searching complexity. The estimation procedure consists of cycles and steps, where a cycle is made of K steps. Each step updates the parameter set of interest for a single user while keeping the other parameter sets constant at their most updated values. Without loss of generality, we follow the natural ordering

 $k = 1, 2, \dots, K$  in updating the users' parameters. Denote

$$\hat{\omega}_{i}^{(k)} = \begin{cases} \left\{ \hat{\xi}_{i}^{(k)}, \, \hat{H}_{i}^{(k)} \right\} & L^{(k)} < M \\ \hat{\xi}_{i}^{(k)}, & L^{(k)} \ge M \end{cases}$$
(27)

as the estimate of  $\omega^{(k)}$  at the *i*th cycle. Define  $\hat{\omega}_{|k,i} = \left\{ \hat{\omega}_{i+1}^{(1)}, \ldots, \hat{\omega}_{i+1}^{(k-1)}, \hat{\omega}_{i}^{(k+1)}, \ldots, \hat{\omega}_{i}^{(K)} \right\}$  and use  $P_V(\tilde{\omega}^{(k)}, \hat{\omega}_{|k,i})$  to indicate the functional dependence of  $P_V$  on  $\left\{ \hat{\omega}_{i+1}^{(1)}, \ldots, \hat{\omega}_{i+1}^{(k-1)}, \tilde{\omega}^{(k)}, \hat{\omega}_{i}^{(k+1)}, \ldots, \hat{\omega}_{i}^{(K)} \right\}$ . At the *k*th step of the (i+1)th cycle, the alternating projection algorithm updates the estimate of  $\omega^{(k)}$  by solving the following minimisation problem [15 eq. (34)–(39)]

$$\hat{\omega}_{i+1}^{(k)} = \arg\min_{\tilde{\omega}^{(k)}, \tilde{X}} \left\| \boldsymbol{P}_{\boldsymbol{B}}^{\perp} \left( \hat{\omega}_{|k,i} \right) \boldsymbol{V}^{(k)} \left( \tilde{\omega}^{(k)} \right) \tilde{\boldsymbol{X}} - \boldsymbol{A} \right\|^2$$
(28)

where  $P_B^{\perp}(\hat{\omega}_{|k,i}) = I_{MK} - B(\hat{\omega}_{|k,i}) [B(\hat{\omega}_{|k,i})]^{\dagger}$  with  $B(\hat{\omega}_{|k,i}) \in \mathbb{C}^{MK \times (\tilde{L}_{sum} - \tilde{L}^{(k)})}$  is a submatrix of  $V(\hat{\omega}^{(k)}, \hat{\omega}_{|k,i})$  formed by deleting the corresponding columns of  $V^{(k)}(\hat{\omega}^{(k)})$ , and  $\tilde{X} \in \mathbb{C}^{\tilde{L}^{(k)} \times PL_s}$  serves as a trial matrix that aims to minimise the above cost function.

According to the value of  $L^{(k)}$ , we can further rewrite (28) into

$$\hat{\omega}_{i+1}^{(k)} = \arg\min_{\tilde{\xi}^{(k)}, \tilde{\boldsymbol{H}}^{(k)}, \tilde{\boldsymbol{X}}} \left\| \boldsymbol{\Pi}_{i}^{(k)} \left( \tilde{\boldsymbol{\xi}}^{(k)} \right) \tilde{\boldsymbol{H}}^{(k)} \tilde{\boldsymbol{X}} - \boldsymbol{A} \right\|^{2}, \quad L^{(k)} < M$$
(29)

$$\hat{\boldsymbol{\xi}}_{i+1}^{(k)} = \arg\min_{\boldsymbol{\tilde{\xi}}^{(k)}, \boldsymbol{\tilde{X}}} \left\| \boldsymbol{\Pi}_{i}^{(k)} (\boldsymbol{\tilde{\xi}}^{(k)}) \boldsymbol{\tilde{X}} - \boldsymbol{A} \right\|^{2}, \quad \boldsymbol{L}^{(k)} \ge \boldsymbol{M} \quad (30)$$

where

$$\boldsymbol{\Pi}_{i}^{(k)}\left(\tilde{\boldsymbol{\xi}}^{(k)}\right) = \boldsymbol{P}_{\boldsymbol{B}}^{\perp}\left(\hat{\boldsymbol{\omega}}_{|k,i}\right) \left(\boldsymbol{v}^{(k)}\left(\tilde{\boldsymbol{\xi}}^{(k)}\right) \otimes \boldsymbol{I}_{M}\right)$$
(31)

First, we consider the case of  $L^{(k)} < M$ . Following the steps similar to [15, eq. (43)–(50)], we know the minimisation problem of (29) in fact falls into the well-known low rank matrix approximation problem. Let SVD of  $\Pi_i^{(k)}(\tilde{\xi}^{(k)})$  be

$$\mathbf{\Pi}_{i}^{(k)}\left(\tilde{\boldsymbol{\xi}}^{(k)}\right) = \boldsymbol{U}_{\Pi,i}^{(k)}\left(\tilde{\boldsymbol{\xi}}^{(k)}\right)\boldsymbol{\Sigma}_{\Pi,i}^{(k)}\left(\tilde{\boldsymbol{\xi}}^{(k)}\right)\left(\boldsymbol{V}_{\Pi,i}^{(k)}\left(\tilde{\boldsymbol{\xi}}^{(k)}\right)\right)^{\mathrm{H}}$$
(32)

where the diagonal entries of  $\Sigma_{\Pi,i}^{(k)}(\tilde{\xi}^{(k)}) \in \mathbb{C}^{M \times M}$  are the non-zero singular values,  $U_{\Pi,i}^{(k)}(\tilde{\xi}^{(k)}) \in \mathbb{C}^{MK \times M}$  and  $V_{\Pi,i}^{(k)}(\tilde{\xi}^{(k)}) \in \mathbb{C}^{M \times M}$  denote the left and right singular vector matrices that correspond to the *M* non-zero singular values, respectively.

Define  $\Xi_i^{(k)}(\tilde{\xi}^{(k)}) = (U_{\Pi,i}^{(k)}(\tilde{\xi}^{(k)}))^H R U_{\Pi,i}^{(k)}(\tilde{\xi}^{(k)})$  where  $R = AA^H$ . Denote the *M* eigenvalues of  $\Xi_i^{(k)}(\tilde{\xi}^{(k)})$  in ascending order and the corresponding eigenvectors by  $\lambda_{l,i}^{(k)}(\tilde{\xi}^{(k)})$  and  $\nu_{l,i}^{(k)}(\tilde{\xi}^{(k)})$ , l = 1, 2, ..., M, respectively. According to [15], the CFO estimation for the *k*th user at

the (i+1)th cycle can be expressed as

$$\hat{\xi}_{i+1}^{(k)} = \arg\max_{\tilde{\xi}^{(k)}} \sum_{l=M-L^{(k)}+1}^{M} \lambda_{l,i}^{(k)} \left(\tilde{\xi}^{(k)}\right)$$
(33)

The channel estimation for the kth user at this step can be then obtained from [15]

$$\hat{\boldsymbol{H}}_{i+1}^{(k)} = \boldsymbol{V}_{\Pi,i}^{(k)} \left( \hat{\boldsymbol{\xi}}_{i+1}^{(k)} \right) \\ \times \left( \boldsymbol{\Sigma}_{\Pi,i}^{(k)} \left( \hat{\boldsymbol{\xi}}_{i+1}^{(k)} \right) \right)^{-1} \left[ \boldsymbol{\nu}_{M-L^{(k)}+1,i}^{(k)} \left( \hat{\boldsymbol{\xi}}_{i+1}^{(k)} \right), \dots, \boldsymbol{\nu}_{M,i}^{(k)} \left( \hat{\boldsymbol{\xi}}_{i+1}^{(k)} \right) \right]$$
(34)

Next, we consider the case of  $L^{(k)} \ge M$ . In this situation, we can rewrite (30) into

$$\hat{\xi}_{i+1}^{(k)} = \arg\min_{\tilde{\xi}^{(k)}} \left\| \left( I_{MK} - \Pi_{i}^{(k)} (\tilde{\xi}^{(k)}) \Pi_{i}^{(k)} (\tilde{\xi}^{(k)}) \right)^{\dagger} \right) A \right\|^{2}$$

$$= \arg\max_{\tilde{\xi}^{(k)}} \left\| \left( U_{\Pi,i}^{(k)} (\tilde{\xi}^{(k)}) \right)^{\mathrm{H}} A \right\|^{2}$$

$$= \arg\max_{\tilde{\xi}^{(k)}} \operatorname{Tr} \left\{ \Xi_{i}^{(k)} (\tilde{\xi}^{(k)}) \right\}$$
(35)

Note that  $\operatorname{Tr}\left\{\Xi_{i}^{(k)}\left(\tilde{\xi}^{(k)}\right)\right\} = \sum_{l=1}^{M} \lambda_{l,i}^{(k)}\left(\tilde{\xi}^{(k)}\right)$ . Thus, combining both (33) and (35), an unified form of the proposed GMLE can be described as follows

$$\hat{\xi}_{i+1}^{(k)} = \arg\max_{\tilde{\xi}^{(k)}} G_i^{(k)} \left(\tilde{\xi}^{(k)}\right)$$
(36)

where the utility function is expressed as

$$G_{i}^{(k)}\left(\tilde{\xi}^{(k)}\right) = \sum_{l=M-\bar{L}^{(k)}+1}^{M} \lambda_{l,i}^{(k)}\left(\tilde{\xi}^{(k)}\right)$$
(37)

#### 5 Performance lower bound of GMLE

In this section, we assume that after the iteration of infinite cycles, to estimate the CFO of the *k*th user using (36), the effect from the parameter estimation error of the other users can be neglected. We then derive the analytical mean square error (MSE) for this simplified situation. Thus, the following analysed MSE can be considered as the performance lower bound of the GMLE.

We define  $\Lambda_i^{(k)}(\xi) = \Pi_i^{(k)}(\xi) \left(\Pi_i^{(k)}(\xi)\right)^{\mathrm{H}} = \boldsymbol{P}_{\boldsymbol{B}}^{\perp}(\hat{\omega}_{|k,i})$   $\Gamma^{(k)}(\xi) \left(\Gamma^{(k)}(\xi)\right)^{\mathrm{H}} \boldsymbol{P}_{\boldsymbol{B}}^{\perp}(\hat{\omega}_{|k,i})$ , where  $\Gamma^{(k)}(\xi) = \boldsymbol{v}^{(k)}(\xi) \otimes \boldsymbol{I}_{M}$ . Denote the *MK* eigenvalues of  $\Lambda_i^{(k)}(\xi)$  in descending order and the corresponding eigenvectors by  $\mu_{l,i}^{(k)}(\xi)$  and  $\boldsymbol{e}_{l,i}^{(k)}(\xi)$ ,  $i = 1, 2, \ldots, MK$ , respectively. Following the proof of Appendix E in [15], we know  $\Lambda_i^{(k)}(\xi)$  has rank of *M*. Then, there holds  $U_{\Pi,i}^{(k)}(\xi) = \left[\boldsymbol{e}_{1,i}^{(k)}(\xi), \boldsymbol{e}_{2,i}^{(k)}(\xi), \ldots, \boldsymbol{e}_{M,i}^{(k)}(\xi)\right]$ , and  $\mu_{l,i}^{(k)}(\xi) = 0$ ,  $l = M + 1, \ldots, MK$ . In the following, we omit the parameterised notation  $(\xi)$  for presentation clarity.

We obtain (see equation at the bottom of the page)

where  $\boldsymbol{D} = (\mathbf{j} 2\pi/K) \operatorname{diag}(0, 1, \dots, K-1) \otimes \boldsymbol{I}_M$ .

From Appendix of this paper, we know the first and second derivatives of  $G_i^{(k)}(\xi)$  with respect to  $\xi$  are given by

$$\frac{\partial G_{i}^{(k)}}{\partial \xi} = \sum_{l=M-\bar{L}^{(k)}+1}^{M} \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \nu_{l,i}^{(k)}$$
(38)

(see (39))

where

$$\mathbb{Q}_{i}^{(k)} = \left(\boldsymbol{U}_{\Pi,i}^{(k)}\right)^{\mathrm{H}} \boldsymbol{R} \Omega_{i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1} + \left(\mu_{i}^{(k)}\right)^{-1} \left(\Omega_{i}^{(k)}\right)^{\mathrm{H}} \boldsymbol{R} \boldsymbol{U}_{\Pi,i}^{(k)}$$
(40)

$$\Omega_i^{(k)} = \left( \boldsymbol{I}_{MK} - \boldsymbol{U}_{\Pi,i}^{(k)} \left( \boldsymbol{U}_{\Pi,i}^{(k)} \right)^{\mathrm{H}} \right) \frac{\partial \Lambda_i^{(k)}}{\partial \xi} \boldsymbol{U}_{\Pi,i}^{(k)} \qquad (41)$$

$$\mu_i^{(k)} = \text{diag} \left\{ \mu_{1,i}^{(k)}, \ \mu_{2,i}^{(k)}, \ \dots, \ \mu_{M,i}^{(k)} \right\}$$
(42)

(see (43))

For the *k*th user, we denote  $\Pi_i^{(k)} = P_B^{\perp}(\omega_{|k,i})$  $(\mathbf{v}^{(k)}(\boldsymbol{\xi}^{(k)}) \otimes \mathbf{I}_M)$  and  $\Lambda_i^{(k)} = \Pi_i^{(k)} (\Pi_i^{(k)})^{\mathrm{H}}$ , where  $\omega_{k,i} = \{\omega^{(1)}, \dots, \omega^{(k-1)}, \omega^{(k+1)}, \dots, \omega^{(K)}\}$  is formed by the perfect knowledge of the parameters of the other users, except the *k*th user. Denote the *M* non-zero eigenvalues of  $\Lambda_i^{(k)}$  in descending order and the corresponding eigenvectors by  $\mu_{l,i}^{(k)}$  and  $\mathbf{e}_{l,i}^{(k)}$ ,  $l = 1, 2, \dots, M$ , respectively. Denote  $U_{\Pi,i}^{(k)} = \left[\mathbf{e}_{1,i}^{(k)}, \mathbf{e}_{2,i}^{(k)}, \dots, \mathbf{e}_{M,i}^{(k)}\right]$  as the matrix formed by the eigenvectors that correspond to the non-zero eigenvalues of  $\Lambda_i^{(k)}$ . The eigenvalues in ascending order and the corresponding eigenvectors of  $(U_{\Pi,i}^{(k)})^{H} R_{i} U_{\Pi,i}^{(k)}$  are expressed by  $\lambda_{l,i}^{(k)}$  and  $\nu_{l,i}^{(k)}$ , l = 1, 2, ..., M, respectively. The lower bound of the MSE of the CFO estimation for the

the lower bound of the MSE of the CFO estimation for the kth user can be expressed as [4]

$$\text{MSE}_{\text{ML, LB}}\left\{\xi^{(k)}\right\} = \frac{E\left\{\left(\partial G_{\infty}^{(k)}/\partial\xi\right)^{2}\right\}}{\left(E\left\{\partial^{2}G_{\infty}^{(k)}/\partial^{2}\xi\right\}\right)^{2}}\bigg|_{\xi=\xi^{(k)}}$$
(44)

where  $G_{\infty}^{(k)}$  stands for the utility function of  $G_i^{(k)}$  after infinite iterative cycles. As mentioned earlier, we ignore the effect of the parameter estimation error from the other users in this section. Thus, there holds

$$\begin{cases} \boldsymbol{U}_{\Pi,\infty}^{(k)} = \boldsymbol{U}_{\Pi,i}^{(k)} \\ \boldsymbol{\mu}_{\infty}^{(k)} = \boldsymbol{\mu}_{i}^{(k)} = \operatorname{diag} \left\{ \boldsymbol{\mu}_{1,i}^{(k)}, \ \boldsymbol{\mu}_{2,i}^{(k)}, \ \dots, \ \boldsymbol{\mu}_{M,i}^{(k)} \right\} \\ \boldsymbol{\Omega}_{\infty}^{(k)} = \boldsymbol{\Omega}_{i}^{(k)} = \left( \boldsymbol{I}_{MK} - \boldsymbol{U}_{\Pi,i}^{(k)} \left( \boldsymbol{U}_{\Pi,i}^{(k)} \right)^{\mathrm{H}} \right) \frac{\partial \boldsymbol{\Lambda}_{i}^{(k)}}{\partial \xi} \boldsymbol{U}_{\Pi,i}^{(k)} \\ \frac{\partial \boldsymbol{\Lambda}_{i}^{(k)}}{\partial \xi} = \boldsymbol{P}_{\boldsymbol{B}}^{\perp} (\boldsymbol{\omega}_{|k,i}) \boldsymbol{D} \boldsymbol{\Gamma}^{(k)} \left( \boldsymbol{\Pi}_{i}^{(k)} \right)^{\mathrm{H}} + \boldsymbol{\Pi}_{i}^{(k)} (\boldsymbol{\Gamma}^{(k)})^{\mathrm{H}} \boldsymbol{D}^{\mathrm{H}} \boldsymbol{P}_{\boldsymbol{B}}^{\perp} (\boldsymbol{\omega}_{|k,i}) \end{cases}$$

We denote  $\mathbf{R} = \mathbf{R}_i + \Delta \mathbf{R}$  and  $\nu_{l,\infty}^{(k)} = \nu_{l,i}^{(k)} + \Delta \nu_{l,\infty}^{(k)}$ , where  $\Delta \mathbf{R}$  and  $\Delta \nu_{l,\infty}^{(k)}$  stand for the corresponding deviations from their expectation. Then, according to (38), we have (see (45) in next page)

Based on the fact that  $\left(\boldsymbol{U}_{\Pi,i}^{(k)}\right)^{\mathrm{H}}\boldsymbol{R}_{i}\Omega_{i}^{(k)} = 0$ , we can simplify  $E\{\left(\partial G_{\infty}^{(k)}/\partial \xi\right)^{2}\}$  into (see (46) and (47) in next page)

The derivation from (46) to (47) is based on the fact of [18, eq. (12)]

$$E\left\{\boldsymbol{x}_{1}^{\mathrm{H}}\Delta\boldsymbol{R}\boldsymbol{x}_{2}\boldsymbol{x}_{3}^{\mathrm{H}}\Delta\boldsymbol{R}\boldsymbol{x}_{4}\right\} = \frac{1}{PL_{s}}\boldsymbol{x}_{1}^{\mathrm{H}}\boldsymbol{R}_{i}\boldsymbol{x}_{4}\boldsymbol{x}_{3}^{\mathrm{H}}\boldsymbol{R}_{i}\boldsymbol{x}_{2} \qquad (48)$$

$$\frac{\partial \Lambda_i^{(k)}}{\partial \xi} = \boldsymbol{P}_{\boldsymbol{B}}^{\perp}(\hat{\boldsymbol{\omega}}_{|k,i})\boldsymbol{D}\Gamma^{(k)}(\boldsymbol{\Pi}_i^{(k)})^{\mathrm{H}} + \boldsymbol{\Pi}_i^{(k)}(\Gamma^{(k)})^{\mathrm{H}}\boldsymbol{D}^{\mathrm{H}}\boldsymbol{P}_{\boldsymbol{B}}^{\perp}(\hat{\boldsymbol{\omega}}_{|k,i})$$
$$\frac{\partial^2 \Lambda_i^{(k)}}{\partial \xi^2} = \boldsymbol{P}_{\boldsymbol{B}}^{\perp}(\hat{\boldsymbol{\omega}}_{|k,i})(2\boldsymbol{D}\Gamma^{(k)}(\Gamma^{(k)})^{\mathrm{H}}\boldsymbol{D}^{\mathrm{H}} + \boldsymbol{D}\boldsymbol{D}\Gamma^{(k)}(\Gamma^{(k)})^{\mathrm{H}} + \Gamma^{(k)}(\Gamma^{(k)})^{\mathrm{H}}\boldsymbol{D}^{\mathrm{H}}\boldsymbol{D}^{\mathrm{H}})\boldsymbol{P}_{\boldsymbol{B}}^{\perp}(\hat{\boldsymbol{\omega}}_{|k,i})$$

$$\frac{\partial^2 G_i^{(k)}}{\partial \xi^2} = \sum_{l=M-\bar{L}^{(k)}+1}^M 2 \Re \left\{ \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \mathbb{M}_i^{(k)} \nu_{l,i}^{(k)} + \sum_{z=1}^{M-\bar{L}^{(k)}} \frac{\left| \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \mathbb{Q}_i^{(k)} \nu_{z,i}^{(k)} \right|^2}{\lambda_{l,i}^{(k)} - \lambda_{z,i}^{(k)}} \right\}$$
(39)

$$\mathbb{M}_{i}^{(k)} = \left(\mu_{i}^{(k)}\right)^{-1} \left(\Omega_{i}^{(k)}\right)^{\mathrm{H}} R \Omega_{i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1} - \Xi_{i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1} \left(\Omega_{i}^{(k)}\right)^{\mathrm{H}} \frac{\partial \Lambda_{i}^{(k)}}{\partial \xi} U_{\Pi,i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1} \\
+ \left(U_{\Pi,i}^{(k)}\right)^{\mathrm{H}} R \left(I_{MK} - U_{\Pi,i}^{(k)} \left(U_{\Pi,i}^{(k)}\right)^{\mathrm{H}}\right) \left(\frac{\partial \Lambda_{i}^{(k)}}{\partial \xi} \Omega_{i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-2} + \frac{\partial^{2} \Lambda_{i}^{(k)}}{\partial \xi^{2}} U_{\Pi,i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1}\right) \\
- 2 \left(U_{\Pi,i}^{(k)}\right)^{\mathrm{H}} R \Omega_{i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1} \left(U_{\Pi,i}^{(k)}\right)^{\mathrm{H}} \frac{\partial \Lambda_{i}^{(k)}}{\partial \xi} U_{\Pi,i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1}$$
(43)

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On the other hand, from (39), we obtain (see (49))

Note that  $\left(\Omega_{i}^{(k)}\right)^{\mathrm{H}}\left(\partial\Lambda_{i}^{(k)}/\partial\xi\right)U_{\Pi,i}^{(k)} = \left(\Omega_{i}^{(k)}\right)^{\mathrm{H}}\Omega_{i}^{(k)}$ . Then, combining both (47) and (49), the MSE lower bound of the CFO estimation for the *k*th user can be finally given by (see (50))

#### 6 Simulations

In this section, we assess the proposed CFO estimation algorithms from computer simulations. The total number of subcarriers is taken as N=64 and are divided into O=4subchannels. The QPSK is adopted. The normalised CFO of each user is randomly generated from -0.35 to 0.35. The root MSE (RMSE) of the normalised CFO estimation is adopted as the figure of merit. M = 8 and  $L_s = 8$  are assumed in the following simulations unless otherwise stated. The estimates of GSSE are taken as the initial values of GMLE. For comparison, we include the CFO estimation scheme of [10], referred to as 'ESPRIT'. For fairness, the multiple receive antenna diversity is exploited in ESPRIT as described in [9, eq. (28)]. The number of users in ESPRIT is three since at least one subchannel should be reserved for null subcarriers in ESPRIT. Note that the results of the estimators of [15] are not included here, since they are equivalent to the proposed GSSE and GMLE in this paper when the multipath numbers of all users equal a same value less than the antenna number.

We start by investigating the convergence rate of the proposed GMLE. In Fig. 1, we depict the CFO estimation performance of GMLE as the cycle number increases. The signal-to-noise ratio (SNR) is taken as 20 dB. The analytical lower bound is also included in this figure by the dotted curves as the benchmark. We denote  $L = [L^{(1)}, L^{(2)}, L^{(3)}, L^{(4)}]$  and include several different cases of L in



Fig. 1 CFO estimation RMSE convergence process of GMLE (SNR = 20 dB)

this figure. Bearing in mind that GMLE estimator is initialised by the GSSE estimator, the results explicitly demonstrate the performance improvement of GMLE introduced by the iterative procedure. We see that, for the included different scenarios, the RMSEs of GMLE quickly decline with first few cycles and almost approach the corresponding analytical lower bounds. These indicate the validity of our method and correctness of the analytical results.

In the second example, we consider  $L^{(1)} = L^{(2)}$  and  $L^{(3)} = L^{(4)}$  without loss of generality. We increase  $L^{(k)}$  (k = 3, 4) from 1 to 16, and show the performance evolution of both GSSE and GMLE in Fig. 2 when SNR equals 20 dB. From these results, we can make the following observations.

$$E\left\{\left(\frac{\partial G_{\infty}^{(k)}}{\partial \xi}\right)^{2}\right\} = E\left\{\left(\sum_{l=M-\bar{L}^{(k)}+1}^{M} \left(\nu_{l,i}^{(k)} + \Delta\nu_{l,\infty}^{(k)}\right)^{\mathrm{H}} \left(\left(\mathbf{U}_{\Pi,i}^{(k)}\right)^{\mathrm{H}} (\boldsymbol{R}_{i} + \Delta\boldsymbol{R}) \Omega_{i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1} ... + \left(\mu_{i}^{(k)}\right)^{-1} \Omega_{i}^{(k)} (\boldsymbol{R}_{i} + \Delta\boldsymbol{R}) U_{\Pi,i}^{(k)} \right) \left(\nu_{l,i}^{(k)} + \Delta\nu_{l,\infty}^{(k)}\right)^{2}\right\}$$
(45)

$$E\left\{\left(\frac{\partial G_{\infty}^{(k)}}{\partial \xi}\right)^{2}\right\} \simeq E\left\{\left(\sum_{l=M-\bar{L}^{(k)}+1}^{M} \left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \left(\left(\boldsymbol{U}_{\Pi,i}^{(k)}\right)^{\mathrm{H}} \Delta \boldsymbol{R} \Omega_{\boldsymbol{i}}^{(k)} \left(\mu_{\boldsymbol{i}}^{(k)}\right)^{-1} + \left(\mu_{\boldsymbol{i}}^{(k)}\right)^{-1} \Omega_{\boldsymbol{i}}^{(k)} \Delta \boldsymbol{R} \boldsymbol{U}_{\Pi,\boldsymbol{i}}^{(k)}\right)\nu_{l,\boldsymbol{i}}^{(k)}\right)^{2}\right\}$$
(46)

$$\simeq \sum_{l=M-\bar{L}^{(k)}+1}^{M} \frac{2\sigma_n^2}{PL_s} \Re \left\{ \lambda_{l,i}^{(k)} \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \left( \mu_i^{(k)} \right)^{-1} \left( \Omega_i^{(k)} \right)^{\mathrm{H}} \Omega_i^{(k)} \left( \mu_i^{(k)} \right)^{-1} \nu_{l,i}^{(k)} \right\}$$
(47)

$$E\left\{\frac{\partial^{2} G_{\infty}^{(k)}}{\partial \xi^{2}}\right\} \simeq -\sum_{l=M-\bar{L}^{(k)}+1}^{M} 2\Re\left\{\lambda_{l,i}^{(k)} \left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \left(\mu_{i}^{(k)}\right)^{-1} \left(\Omega_{i}^{(k)}\right)^{\mathrm{H}} \frac{\partial\Lambda_{i}^{(k)}}{\partial\xi} U_{\Pi,i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1} \nu_{l,i}^{(k)}\right\}$$
(49)

$$MSE_{ML, LB} \{\xi^{(k)}\} = \frac{(\sigma_n^2/2PL_s)}{\sum_{l=M-\bar{L}^{(k)}+1}^M \lambda_{l,i}^{(k)} (\nu_{l,i}^{(k)})^H (\mu_i^{(k)})^{-1} (\Omega_i^{(k)})^H \Omega_i^{(k)} (\mu_i^{(k)})^{-1} \nu_{l,i}^{(k)}}$$
(50)



**Fig. 2** *CFO estimation RMSE performance of GSSE and GMLE against*  $L^{(k)}$ , k = 3, 4 (SNR = 20 dB)

First, as expected, ESPRIT has the advantage of insensitivity to the value of  $L^{(k)}$ , whereas both GSSE and GMLE suffer from the performance degradation as the multipath number increases when  $L^{(k)} \leq 8$ , k=3, 4, which coincides with the observations in [15]. It is seen that, GMLE always behaves better than GSSE when  $L^{(k)}$  (k=3, 4) are increased from 1 to 16, especially when  $L^{(k)}$  (k=3, 4) are around 8. Moreover, we observe that the performance of GMLE is gradually degraded when  $L^{(k)}$  is increased from 1 to 8 and is basically unchanged when  $L^{(k)}$  becomes even larger. Thus, this indicates GMLE also has the advantage of being more insensitive to  $L^{(k)}$  as compared with GSSE.

Second and more interestingly, we can observe that GSSE behaves worst when  $L^{(k)}$  (k=3, 4) equal 8 and its performance can be improved when  $L^{(k)}$  (k=3, 4) become larger than 8. This may be explained as follows. When  $L^{(k)} \ge M$ , the increase of multipath number would improve the non-singularity of the correlation matrices  $R_{\bar{z}\bar{z}}^{(k)}$  statistically, k=3, 4, which is beneficial to the estimation performance of GSSE.

Third, it is seen that, the simulation results of GMLE also match the corresponding analytical curves of lower bound, which indicates the correctness of theoretical analysis.



**Fig. 3** *CFO RMSE performance comparison between GSSE and GMLE as a function of SNR* 



**Fig. 4** *CFO RMSE performance comparison between GSSE and GMLE as a function of*  $L_s$ 

In Fig. 3, we show the performance comparison between GSSE and GMLE as a function of SNR with different configurations of multipath numbers. The comparison clearly demonstrates the superiority of GMLE as compared with GSSE. We also see that ESPRIT achieves better performance than both GSSE and GMLE in the investigated scenarios. However, we should note that the system is fully loaded in both GSSE and GMLE, whereas one subchannel should be reserved in ESPRIT. Thus, our estimators support higher bandwidth efficiency as compared with ESPRIT.

Next, we show the performance comparison among the different estimators as  $L_s$  increases from 2 to 8 in Fig. 4. The SNR is assumed to be 20 dB in this example. The comparison again indicates the superiority of GMLE as compared with GSSE. It is also seen that both GMLE and ESPRIT suffer from only slight performance degradation when fewer block durations are available.

Fig. 5 shows the QPSK and 16QAM symbol error rate (SER) performance comparison between the proposed GMLE and ESPRIT estimator. We assume the multi-antenna employs maximal-ratio receiver the combining decoder after the CFO compensation. The channel responses are assumed to be known to the receiver.



Fig. 5 SER performance comparison as a function of SNR

We consider two block durations are adopted in this example. The results with perfect CFO estimation are also plotted as the benchmark, referred to as 'genie-aided' in this figure. We can observe the performance gap between our method and the genie-aided results. However, it is seen that the proposed method can achieve comparable SER performance with the ESPRIT estimator.

#### 7 Conclusions

We have developed two generalised blind CFO estimators for the interleaved OFDMA uplink with multi-antenna receiver. We consider the more general scenarios that the number of multipaths from some users may be larger than the number of antennas at the receiver. We have found that, as long as the numbers of multipaths from two of the users are smaller than the number of antennas at the receiver, the proposed estimators support fully loaded transmissions. Both the theoretical and numeral results are provided, which corroborate the proposed studies.

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#### 9 References

- Wong, C.Y., Cheng, R.S., Lataief, K.B., Murch, R.D.: 'Multiuser OFDM with adaptive subcarrier, bit, and power allocation', *IEEE J. Sel. Areas Commun.*, 1999, **17**, (10), pp. 1747–1758
- 2 Fa, R., Zhang, L.: 'Generalised grouped minimum mean-squared error-based multi-stage interference cancellation scheme for orthogonal frequency division multiple access uplink systems with carrier frequency offsets', *IET Commun.*, 2013, 7, (7), pp. 685–695
- 3 Tureli, U., Liu, H., Zoltowski, M.D.: 'OFDM blind carrier frequency offset estimation: ESPRIT', *IEEE Trans. Commun.*, 2001, 48, (9), pp. 1459–1461
- Zhang, W., Yin, Q.: 'Blind maximum likelihood carrier frequency offset estimation for OFDM with multi-antenna receiver', *IEEE Trans. Signal Process.*, 2013, **61**, (9), pp. 2295–2307
  Sun, P., Zhang, L.: 'Low complexity pilot aided frequency
- 5 Sun, P., Zhang, L.: 'Low complexity pilot aided frequency synchronization for OFDMA uplink transmission', *IEEE Trans. Wirel. Commun.*, 2009, 8, (7), pp. 3758–3769
- 6 Wang, Z., Xin, Y., Mathew, G.: 'Iterative carrier-frequency offset estimation for generalized OFDMA uplink transmission', *IEEE Trans. Wirel. Commun.*, 2009, 8, (3), pp. 1373–1383
- 7 Barbarossa, S., Pompili, M., Giannakis, G.B.: 'Channel-independent synchronization of orthogonal frequency division multiple access systems', *IEEE J. Sel. Areas Commun.*, 2002, **20**, (2), pp. 474–486
- 8 Yao, Y., Giannakis, G.B.: 'Blind carrier frequency offset estimation in SISO, MIMO, and multiuser OFDM systems', *IEEE Trans. Commun.*, 2005, **53**, (1), pp. 173–183
- 9 Cao, Z., Tureli, U., Yao, Y.-D.: 'Deterministic multiuser carrier-frequency offset estimation for interleaved OFDMA uplink', *IEEE Trans. Commun.*, 2004, **52**, (9), pp. 1585–1594
- Lee, J., Lee, S., Band, K.-J., Cha, S., Hong, D.: 'Carrier frequency offset estimation using ESPRIT for interleaved OFDMA uplink systems', *IEEE Trans. Veh. Tech.*, 2007, 56, (9), pp. 3227–3231
- 11 Hisieh, H.T., Wu, W.R.: 'Blind maximum-likelihood carrier-frequency offset estimation for interleaved OFDMA uplink systems', *IEEE Trans. Veh. Technol.*, 2011, **60**, (1), pp. 160–173
- 12 Zhu, Y., Letaief, K.B.: 'CFO estimation and compensation in single carrier interleaved FDMA systems'. Proc. IEEE GLOBECOM, Honolulu, USA, November 2009
- 13 Wu, K.H., Fang, W.H., Chen, Y.T.: 'Joint carrier frequency offset and direction of arrival estimation via hierarchical ESPRIT for interleaved OFDMA/SDMA uplink systems'. Proc. IEEE VTC Spring, Taipei, Taiwan, May 2010

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- 14 Du, R., Wang, J., Liu, F.: 'Unitary-ESPRIT algorithm for carrier frequency offset estimation for interleaved OFDMA uplink systems', *Wirel. Pers. Commun.*, 2013, 69, pp. 1615–1627
- 15 Zhang, W., Gao, F., Yin, Q., Nallanathan, A.: 'Blind carrier frequency offset estimation for interleaved OFDMA uplink', *IEEE Trans. Signal Process.*, 2012, **60**, (7), pp. 3616–3627
- 16 Schmidt, R.O.: 'Multiple emitter location and signal parameter estimation'. Proc. RADC Spectral Estimation Workshop, New York, 1979, pp. 243–258
- 17 Ziskind, I., Wax, M.: 'Maximum likelihood localization of multiple sources by alternating projection', *IEEE Trans. Acoust. Speech Signal Process.*, 1988, 36, (10), pp. 1553–1560
- 18 Rao, B.D., Hari, K.V.S.: 'Effect of spatial smoothing on the performance of MUSIC and the minimum-norm method', *IEE Proc.*, 1990, **137**, (6), pp. 449–458

#### 10 Appendix

In the following, we omit the parameterised notation ( $\xi$ ) for presentation clarity. The first derivative of  $G_i^{(k)}$  with respect to  $\xi$  is expressed as

$$\frac{\partial G_i^{(k)}}{\partial \xi} = \sum_{l=M-\bar{L}^{(k)}+1}^M \left(\nu_{l,i}^{(k)}\right)^H \frac{\partial \Xi_i^{(k)}}{\partial \xi} \nu_{l,i}^{(k)}$$
(51)

where

$$\frac{\partial \Xi_{i}^{(k)}}{\partial \xi} = \left(\frac{\partial U_{\Pi,i}^{(k)}}{\partial \xi}\right)^{\mathrm{H}} R U_{\Pi,i}^{(k)} + \left(U_{\Pi,i}^{(k)}\right)^{\mathrm{H}} R \frac{\partial U_{\Pi,i}^{(k)}}{\partial \xi}$$

We have

$$\frac{\partial \boldsymbol{e}_{l,i}^{(k)}}{\partial \xi} = \sum_{j \neq l}^{M} \boldsymbol{e}_{j,i}^{(k)} \frac{\left(\boldsymbol{e}_{j,i}^{(k)}\right)^{\mathrm{H}} \left(\partial \Lambda_{i}^{(k)} / \partial \xi\right) \boldsymbol{e}_{l,i}^{(k)}}{\mu_{l,i}^{(k)} - \mu_{j,i}^{(k)}} + \left(\boldsymbol{I}_{MK} - \boldsymbol{U}_{\Pi,i}^{(k)} \left(\boldsymbol{U}_{\Pi,i}^{(k)}\right)^{\mathrm{H}}\right) \frac{\left(\partial \Lambda_{i}^{(k)} / \partial \xi\right) \boldsymbol{e}_{l,i}^{(k)}}{\mu_{l,i}^{(k)}}$$
(52)

Consequently, we have

$$rac{\partial oldsymbol{U}_{\Pi,i}^{(k)}}{\partial \xi} = oldsymbol{U}_{\Pi,i}^{(k)} oldsymbol{E}_i^{(k)} + \Omega_i^{(k)} \Big( oldsymbol{\mu}_i^{(k)} \Big)^{-1}$$

where  $\mu_i^{(k)}$  is defined in (43) and  $E_i^{(k)} \in \mathbb{C}^{M \times M}$  is constructed as follows. We have

$$\boldsymbol{E}_{i}^{(k)}(l, j) = \frac{\left(\boldsymbol{e}_{j,i}^{(k)}\right)^{\mathrm{H}} \left(\partial \Lambda_{i}^{(k)} / \partial \xi\right) \boldsymbol{e}_{l,i}^{(k)}}{\lambda_{j,i}^{(k)} - \lambda_{l,i}^{(k)}}$$

when  $l \neq j$ , and  $E_i^{(k)}(l, l) = 0, l = 1, 2, \dots, M$ . Hence, (51) can be rewritten as (see (53) in next page)

Bearing in mind that  $E_i^{(k)} + (E_i^{(k)})^H = 0$ , we simplify (53) into

$$\frac{\partial G_i^{(k)}}{\partial \xi} = \sum_{l=M-\bar{L}^{(k)}+1}^M \left( \nu_{l,i}^{(k)} \right)^H \mathbb{Q}_i^{(k)} \nu_{l,i}^{(k)}$$
(54)

where  $\mathbb{Q}_{i}^{(k)} = \Sigma_{i}^{(k)} (\mu_{i}^{(k)})^{-1} + (\mu_{i}^{(k)})^{-1} (\Sigma_{i}^{(k)})^{\mathrm{H}}$  and  $\Sigma_{i}^{(k)} = (U_{\Pi,i}^{(k)})^{\mathrm{H}} R\Omega_{i}^{(k)}$ . Then, we arrive at (38).

Next, based on (51), the second derivative of  $G_i^{(k)}$  can be given by

$$\frac{\partial^{2} G_{i}^{(k)}}{\partial \xi^{2}} = \sum_{l=M-\bar{L}^{(k)}+1}^{M} \left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \frac{\partial \mathbb{Q}_{i}^{(k)}}{\partial \xi} \nu_{l,i}^{(k)} + \left(\nu_{l,i}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \frac{\partial \nu_{l,i}^{(k)}}{\partial \xi} + \left(\frac{\partial \nu_{l,i}^{(k)}}{\partial \xi}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \nu_{l,i}^{(k)}$$
(55)

We obtain (see (56))

where

$$\frac{\partial \Sigma_{i}^{(k)}}{\partial \xi} = \left(\frac{\partial \left(\boldsymbol{U}_{\Pi,i}^{(k)}\right)^{\mathrm{H}}}{\partial \xi}\right)^{\mathrm{H}} \boldsymbol{R} \Omega_{i}^{(k)} + \left(\boldsymbol{U}_{\Pi,i}^{(k)}\right)^{\mathrm{H}} \mathbf{R} \frac{\partial \Omega_{i}^{(k)}}{\partial \xi}$$
$$= \mathbb{H}_{i}^{(k)} + \Sigma_{i}^{(k)} \boldsymbol{E}_{i}^{(k)} + \left(\boldsymbol{E}_{i}^{(k)}\right)^{\mathrm{H}} \Sigma_{i}^{(k)} \tag{57}$$

(see (58 and 59))

By substituting (57) into (56), we have (see (60))

where

$$\frac{\partial (\mu_i^{(k)})^{-1}}{\partial \xi} = -((\mu_i^{(k)})^{-1} (U_{\Pi,i}^{(k)})^H \frac{\partial \Lambda_i^{(k)}}{\partial \xi} U_{\Pi,i}^{(k)} (\mu_i^{(k)})^{-1}) \circ I_M$$

Moreover, note that (see equation at the bottom of the page) Then, there holds

$$\left(\boldsymbol{\nu}_{l,i}^{(k)}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \frac{\partial \boldsymbol{\nu}_{l,i}^{(k)}}{\partial \xi} = \sum_{z \neq l}^{M} \left(\boldsymbol{\nu}_{l,i}^{(k)}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \boldsymbol{\nu}_{z,i}^{(k)} \left(\boldsymbol{\nu}_{z,i}^{(k)}\right)^{\mathrm{H}} \left(\boldsymbol{E}_{i}^{(k)}\right)^{\mathrm{H}} \boldsymbol{\nu}_{l,i}^{(k)} + \frac{\left| (\boldsymbol{\nu}_{l,i}^{(k)})^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \boldsymbol{\nu}_{z,i}^{(k)} \right|^{2}}{\lambda_{l,i}^{(k)} - \lambda_{z,i}^{(k)}}$$

$$(61)$$

On the other side, we obtain (see (62) in next page) Substituting both (60) and (62) into (55), we arrive at

$$\frac{\partial G_{i}^{(k)}}{\partial \xi} = \sum_{l=M-\bar{L}^{(k)}+1}^{M} \left( \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \left( U_{\Pi,i}^{(k)} \right)^{\mathrm{H}} R \Omega_{i}^{(k)} \left( \mu_{i}^{(k)} \right)^{-1} \nu_{l,i}^{(k)} + \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \left( \mu_{i}^{(k)} \right)^{-1} \left( \Omega_{i}^{(k)} \right)^{\mathrm{H}} R U_{\Pi,i}^{(k)} \nu_{l,i}^{(k)} + \lambda_{l,i}^{(k)} \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \left( E_{i}^{(k)} + \left( E_{i}^{(k)} \right)^{\mathrm{H}} \right) \nu_{l,i}^{(k)}$$
(53)

$$\left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \frac{\partial \mathbb{Q}_{i}^{(k)}}{\partial \xi} \nu_{l,i}^{(k)} = 2 \,\Re \left\{ \left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \left(\frac{\partial \Sigma_{i}^{(k)}}{\partial \xi}\right) \left(\mu_{i}^{(k)}\right)^{-1} \nu_{l,i}^{(k)} + \nu_{l,i}^{(k)} \Sigma_{i}^{(k)} \frac{\partial \left(\mu_{i}^{(k)}\right)^{-1}}{\partial \xi} \nu_{l,i}^{(k)} \right\}$$
(56)

$$\mathbb{H}_{i}^{(k)} = \left(\mu_{i}^{(k)}\right)^{-1} \left(\Omega_{i}^{(k)}\right)^{\mathrm{H}} R \Omega_{i}^{(k)} + \left(U_{\Pi,i}^{(k)}\right)^{\mathrm{H}} R \left(I - U_{\Pi,i}^{(k)} \left(U_{\Pi,i}^{(k)}\right)^{\mathrm{H}}\right) \left(\frac{\partial \Lambda_{i}^{(k)}}{\partial \xi} \Omega_{i}^{(k)} (\mu_{i}^{(k)})^{-1} + \frac{\partial^{2} \Lambda_{i}^{(k)}}{\partial \xi^{2}} U_{\Pi,i}^{(k)}\right) \\
- \left(\Sigma_{i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1} \left(U_{\Pi,i}^{(k)}\right)^{\mathrm{H}} + \Xi_{i}^{(k)} (\mu_{i}^{(k)})^{-1} \left(\Omega_{i}^{(k)}\right)^{\mathrm{H}}\right) \frac{\partial \Lambda_{i}^{(k)}}{\partial \xi} U_{\Pi,i}^{(k)} \tag{58}$$

$$\frac{\partial^2 \Lambda_i^{(k)}}{\partial \xi^2} = \boldsymbol{P}_{\boldsymbol{B}}^{\perp} \left( \hat{\boldsymbol{\omega}}_{|k,i} \right) \left( 2D\Gamma^{(k)} (\Gamma^{(k)})^{\mathrm{H}} D^{\mathrm{H}} + DD\Gamma^{(k)} (\Gamma^{(k)})^{\mathrm{H}} + \Gamma^{(k)} (\Gamma^{(k)})^{\mathrm{H}} D^{\mathrm{H}} D^{\mathrm{H}} D^{\mathrm{H}} \right) \boldsymbol{P}_{\boldsymbol{B}}^{\perp} \left( \hat{\boldsymbol{\omega}}_{|k,i} \right)$$
(59)

$$\left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \frac{\partial \mathbb{Q}_{i}^{(k)}}{\partial \xi} \nu_{l,i}^{(k)} = 2 \,\Re \left\{ \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \mathbb{H}_{i}^{(k)} (\mu_{i}^{(k)})^{-1} \nu_{l,i}^{(k)} + \nu_{l,i}^{(k)} \Sigma_{i}^{(k)} \frac{\partial \left( \mu_{i}^{(k)} \right)^{-1}}{\partial \xi} \nu_{l,i}^{(k)} \right\} + 2 \,\Re \left\{ \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \left( \Sigma_{i}^{(k)} \boldsymbol{E}_{i}^{(k)} + \left( \boldsymbol{E}_{i}^{(k)} \right)^{\mathrm{H}} \Sigma_{i}^{(k)} \right) \left( \mu_{i}^{(k)} \right)^{-1} \nu_{l,i}^{(k)} \right\}$$
(60)

$$\begin{pmatrix} \nu_{l,i}^{(k)} \end{pmatrix}^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \frac{\partial \nu_{l,i}^{(k)}}{\partial \xi} = \sum_{z \neq l}^{M} \begin{pmatrix} \nu_{l,i}^{(k)} \end{pmatrix}^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \nu_{z,i}^{(k)} \frac{\left(\nu_{z,i}^{(k)}\right)^{\mathrm{H}} \left(\partial \Xi_{i}^{(k)} / \partial \xi\right) \nu_{l,i}^{(k)}}{\lambda_{l,i}^{(k)} - \lambda_{z,i}^{(k)}} \\ \left(\nu_{z}^{(k)}\right)^{\mathrm{H}} \frac{\partial \Xi_{i}^{(k)}}{\partial \xi} \nu_{l,i}^{(k)} = \left(\lambda_{l,i}^{(k)} - \lambda_{z,i}^{(k)}\right) \left(\nu_{z,i}^{(k)}\right)^{\mathrm{H}} \left(\boldsymbol{E}_{i}^{(k)}\right)^{\mathrm{H}} \nu_{l,i}^{(k)} + \left(\nu_{z,i}^{(k)}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \nu_{l,i}^{(k)}$$

(see (63))

Note that (see equation at the bottom of the page)

where

$$\boldsymbol{M}_{i}^{(k)} = \left( \left( \boldsymbol{\mu}_{i}^{(k)} \right)^{-1} \left( \boldsymbol{U}_{\Pi,i}^{(k)} \right)^{\mathrm{H}} \left( \partial \boldsymbol{\Lambda}_{i}^{(k)} / \partial \boldsymbol{\xi} \right) \boldsymbol{U}_{\Pi,i}^{(k)} \left( \boldsymbol{\mu}_{i}^{(k)} \right)^{-1} \right)$$
  
$$\circ \left( \boldsymbol{1}_{M \times 1} - \boldsymbol{I}_{M} \right)$$

Then, we can rewrite  $\left(\partial^2 G_i^{(k)}/\partial \xi^2\right)$  as (see (64))

By substituting  $\mathbb{H}_{i}^{(k)}$  from (58) into (64), we can arrive at (39) after some simple manipulations. This completes the proof.

$$\begin{split} \sum_{l=M-\bar{L}^{(k)}+1}^{M} \left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \frac{\partial \nu_{l,i}^{(k)}}{\partial \xi} + \left(\frac{\partial \nu_{l,i}^{(k)}}{\partial \xi}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \nu_{l,i}^{(k)} \\ &= \sum_{l=M-\bar{L}^{(k)}+1}^{M} \left[\sum_{z=1}^{M-\bar{L}^{(k)}} 2 \frac{\left|\left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \nu_{z,i}^{(k)}\right|^{2}}{\lambda_{l,i}^{(k)} - \lambda_{z,i}^{(k)}} + 2 \Re\left\{\left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \left(E_{i}^{(k)}\right)^{\mathrm{H}} \nu_{l,i}^{(k)}\right\}\right] \end{split}$$
(62)  
$$\frac{\partial^{2} G_{i}^{(k)}}{\partial \xi^{2}} = \sum_{l=M-\bar{L}^{(k)}+1}^{M} 2 \Re\left\{\left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \mathbb{H}_{i}^{(k)} \left(\mu_{i}^{(k)}\right)^{-1} \nu_{l,i}^{(k)} + \left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \Sigma_{i}^{(k)} \frac{\partial\left(\mu_{i}^{(k)}\right)^{-1}}{\partial \xi} \nu_{l,i}^{(k)}\right\} \\ &+ 2 \Re\left\{\left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \left(\Sigma_{i}^{(k)} E_{i}^{(k)} + \left(E_{i}^{(k)}\right)^{\mathrm{H}} \Sigma_{i}^{(k)}\right) \left(\mu_{i}^{(k)}\right)^{-1} \nu_{l,i}^{(k)} + \left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \left(E_{i}^{(k)}\right)^{\mathrm{H}} \nu_{l,i}^{(k)}\right\} \\ &+ \sum_{z=1}^{M-\bar{L}^{(k)}} 2 \frac{\left|\left(\nu_{l,i}^{(k)}\right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \nu_{z,i}^{(k)}\right|^{2}}{\lambda_{l,i}^{(k)} - \lambda_{z,i}^{(k)}} \end{aligned}$$

$$2 \Re \left\{ \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \left( \Sigma_{i}^{(k)} \boldsymbol{E}_{i}^{(k)} + \left( \boldsymbol{E}_{i}^{(k)} \right)^{\mathrm{H}} \Sigma_{i}^{(k)} \right) \left( \mu_{i}^{(k)} \right)^{-1} \nu_{l,i}^{(k)} + \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \left( \boldsymbol{E}_{i}^{(k)} \right)^{\mathrm{H}} \nu_{l,i}^{(k)} \right\}$$
$$= 2 \Re \left\{ \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \Sigma_{i}^{(k)} \left( \boldsymbol{E}_{i}^{(k)} \left( \mu_{i}^{(k)} \right)^{-1} - \left( \mu_{i}^{(k)} \right)^{-1} \boldsymbol{E}_{i}^{(k)} \right) \nu_{l,i}^{(k)} \right\}$$
$$= -2 \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \Sigma_{i}^{(k)} \boldsymbol{M}_{i}^{(k)} \nu_{l,i}^{(k)}$$

$$\frac{\partial^{2} G_{i}^{(k)}}{\partial \xi^{2}} = \sum_{l=M-\bar{L}^{(k)}+1}^{M} 2 \Re \left\{ \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \left( \mathbb{H}_{i}^{(k)} \left( \mu_{i}^{(k)} \right)^{-1} - \Sigma_{i}^{(k)} \left( \mu_{i}^{(k)} \right)^{-1} \left( U_{\Pi,i}^{(k)} \right)^{\mathrm{H}} \frac{\partial \Lambda_{i}^{(k)}}{\partial \xi} U_{\Pi,i}^{(k)} \left( \mu_{i}^{(k)} \right)^{-1} \nu_{l,i}^{(k)} \right\} + \sum_{z=1}^{M-\bar{L}^{(k)}} 2 \frac{\left| \left( \left( \nu_{l,i}^{(k)} \right)^{\mathrm{H}} \mathbb{Q}_{i}^{(k)} \nu_{z,i}^{(k)} \right|^{2}}{\lambda_{l,i}^{(k)} - \lambda_{z,i}^{(k)}} \right|^{2} \tag{64}$$

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